

1) per simulare correttamente un segnale bipolare in ingresso, poiché l'opamp è alimentato tra 0 e +10V, devo trovare il valore in DC dell'uscita a metà dinamica

⇓

$$V_{out}|_{DC} = 5V = +10V * \left[\frac{R_b}{R_a + R_b} \right] * \left(1 + \frac{R_2}{R_1} \right)$$

⇓

$$\frac{R_b}{R_a + R_b} = \frac{1}{1 + R_a/R_b} = \frac{5V}{10V} * \frac{1}{1 + R_2/R_1} = \frac{1}{2} * \frac{1}{1 + 10} = \frac{1}{22}$$

$$\hookrightarrow 1 + \frac{R_a}{R_b} = 22 \rightarrow \frac{R_a}{R_b} = 21$$

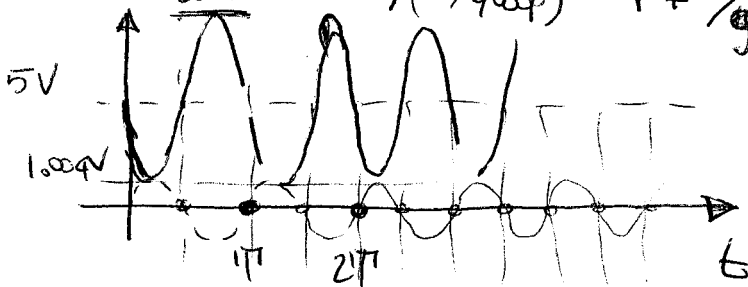
Inoltre per avere $I_{div} \geq 100\mu A \Rightarrow R_a + R_b \leq \frac{+10V}{I_{div}} = \frac{10V}{100\mu A} = 100k$

↳ ad esempio $R_b = 5k$ e $R_a = 21 * 5k = 105k \Rightarrow R_a + R_b = 110k$

② $G_{id} = -\frac{R_2}{R_1} = -10$

$$G_{loop} = -\frac{R_1}{R_1 + R_2} * A_o = -\frac{2.2k}{2.2k + 22k} * 10^4 \approx -909$$

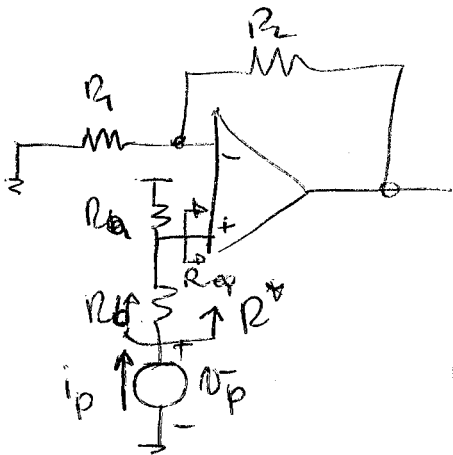
$$\hookrightarrow G_{redc} = G_{id} / (1 - 1/G_{loop}) = \frac{-10}{1 + 1/909} \approx -9.99$$



$$T = \frac{2\pi}{\omega} = \frac{2\pi}{1 \text{krad/s}} = 6.28 \text{ms}$$

$$\Delta V_{out}|_{max} = |400 \text{mV} * (-9.99)| = 3996$$

③



$$R_{id}^* = R_b + R_a // \infty$$

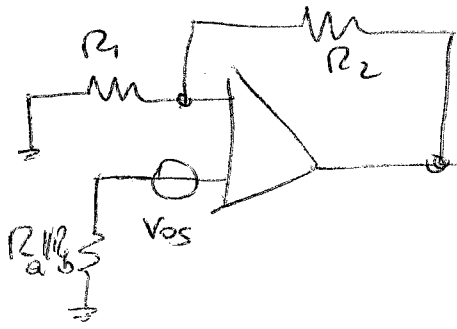
$$\hookrightarrow R^* = R_b + [R_a // (R_{eq})]$$

$$R_{eq} = \frac{R_{eq}^0}{1 - G_{loop}^*} = \frac{2k}{2k} = 4.55 \Omega$$

$$R_{eq}^0 = R_{i0} + R_1 // R_2 = 5 \mu\Omega + 2.2k // 22k = 5.002 \mu\Omega$$

$$G_{loop}^* = -\frac{R_1 // R_{i0}}{R_1 // R_{i0} + R_2} A_o = -909.1$$

4



$$V_{out}|_{V_{os}} = \pm V_{os} \left(1 + \frac{R_2}{R_1} \right) = \pm 10 \text{ mV} \left(1 + \frac{22 \text{ k}\Omega}{2.2 \text{ k}\Omega} \right) = \pm 110 \text{ mV}$$

Si è ancora possibile "simplificare" correttamente il segnale di cui al punto 4 perché la tensione di offset V_{os} ha il valor medio della tensione di uscita al poi di ± 110

5) Per compensare le correnti di bias la resistenza vista in DC dal morsetto non invertente deve uguagliare quello visto dal morsetto invertente.

$$R_1 \parallel R_2 = R_a \parallel R_b$$

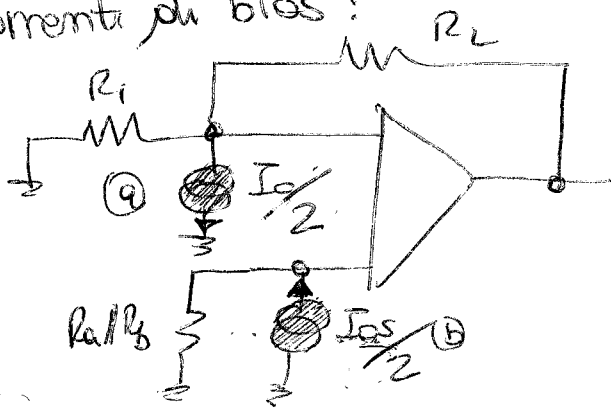
$$= 2 \text{ k}\Omega \Rightarrow R_a \parallel R_b = 2 \text{ k}\Omega \quad \text{ed, inoltre per i requisiti}$$

punto 4) $\frac{R_a}{R_b} = 21 \Rightarrow \left\{ \begin{array}{l} \frac{R_a \cdot R_b}{R_a + R_b} = 2 \text{ k}\Omega \\ \frac{R_a}{R_b} = 21 \end{array} \right. \Rightarrow \frac{21 R_b}{22 R_b} = 2 \text{ k}\Omega$

$$I_{div} = \frac{10 \text{ V}}{R_a + R_b} = \leftarrow R_b = 2.1 \text{ k}\Omega$$

$$= 217 \mu\text{A} \quad R_a = 44 \text{ k}\Omega$$

A questo punto resta solo l'effetto dell'offset delle correnti di bias:

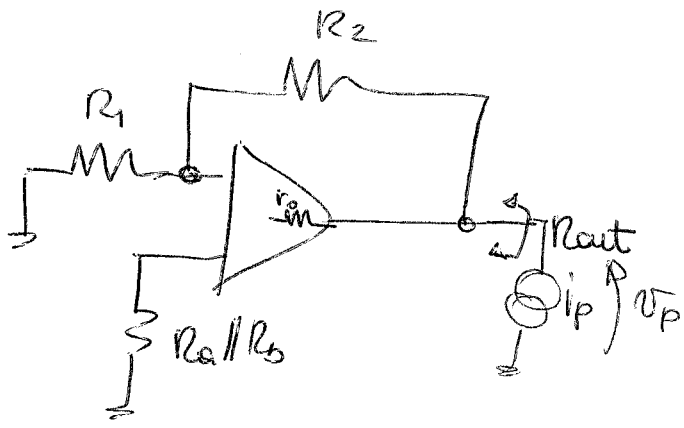


(a) $V_{out} = \frac{\pm I_{os}}{2} \times R_2 = \frac{100 \mu\text{A}}{2} \times 22 \text{ k}\Omega = \pm 1.1 \text{ mV}$

(b) $= \frac{I_{os}}{2} (R_a \parallel R_b) \left(1 + \frac{R_2}{R_1} \right) = \frac{\pm 100 \mu\text{A}}{2} (2 \text{ k}\Omega) (1 + 10) = \pm 1.1 \text{ mV}$

$$\Downarrow V_{o}|_{I_{os}} = \pm 2.2 \text{ mV}$$

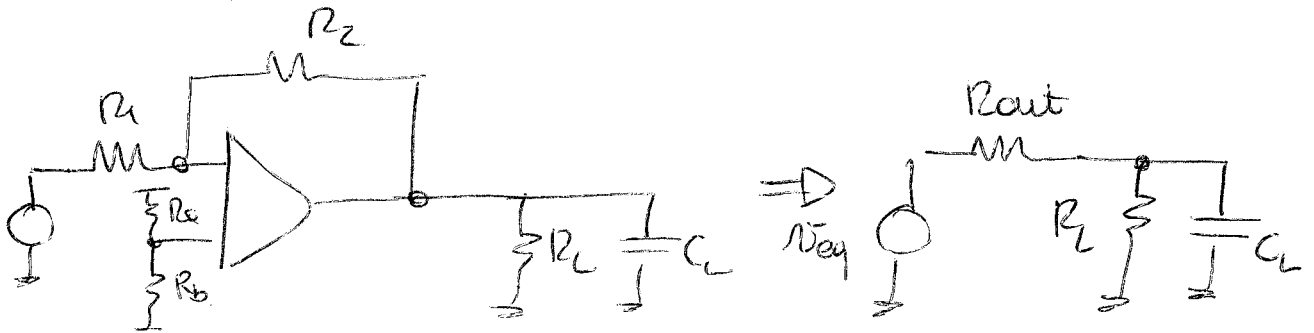
6



$$R_{out} = \frac{R_{out}^o}{1 - G_{loop}^{**}} = \frac{r_o \parallel (R_1 + R_2)}{1 + 907} \approx \frac{47}{908} = 0.052 \Omega$$

$$G_{loop}^{**} = -\frac{R_1}{R_1 + R_2 + r_o} A_o = -\frac{2.2k}{22k + 22k + 47\Omega} \times 10^4 = -907$$

7



$$\tau_p = C_L * (R_L \parallel R_{out}) = 200pF * (10k \parallel 0.052\Omega) = 10ns$$

$$f_p = \frac{1}{2\pi\tau_p} = 15.3 GHz !! \text{ la capacit\`a } C_L \text{, quasi pesa!}$$